## Control Systems And Their Components (EE391)

Lec. 12: Combined state feedback and state estimator

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## Lecture Outline

- Combining regulator with state estimator
- Combining state feedback + state estimator + pre-scaling reference input


## Combining regulator with state estimator

- Now we will study the performance when combining a regulator designed as $u=-\mathbf{K x}$ but implemented as $u=-\mathbf{K} \hat{\mathbf{x}}$

$$
\begin{aligned}
\dot{\mathbf{x}}(t) & =\mathbf{A x}(t)+\mathbf{B} u(t) \quad \Leftarrow \text { plant state equation } \\
& =\mathbf{A x}(t)-\mathbf{B K} \hat{\mathbf{x}}(t) \quad \Leftarrow \text { apply state feedback } u=-\mathbf{K} \hat{\mathbf{x}} \\
& =\mathbf{A x}(t)-\mathbf{B K}(\mathbf{x}(t)-\mathbf{e}(t)) \\
& =(\mathbf{A}-\mathbf{B K}) \mathbf{x}(t)+\mathbf{B K e}(t) \\
\dot{\hat{\mathbf{x}}}(t) & =\mathbf{A} \hat{\mathbf{x}}(t)+\mathbf{B} u(t)+\mathbf{L C}(\mathbf{x}(t)-\hat{\mathbf{x}}(t)) \\
\dot{\mathbf{e}}(t) & =(\mathbf{A}-\mathbf{L C} \mathbf{C} \mathbf{e}(t) \\
{\left[\begin{array}{l}
\dot{\mathbf{x}}(t) \\
\dot{\mathbf{e}}(t)
\end{array}\right] } & =\left[\begin{array}{cc}
\mathbf{A}-\mathbf{B K} & \mathbf{B K} \\
0 & \mathbf{A}-\mathbf{L} \mathbf{C}
\end{array}\right]\left[\begin{array}{l}
\mathbf{x}(t) \\
\mathbf{e}(t)
\end{array}\right] \Leftarrow \text { augmented equation } \\
y(t) & =\left[\begin{array}{ll}
\mathbf{C} & 0
\end{array}\right]\left[\begin{array}{l}
\mathbf{x}(t) \\
\mathbf{e}(t)
\end{array}\right]
\end{aligned}
$$

## Combining regulator with state estimator

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$$
\left[\begin{array}{c}
\dot{\mathbf{x}}(t) \\
\dot{\mathbf{e}}(t)
\end{array}\right]=\underbrace{\left[\begin{array}{cc}
\mathbf{A}-\mathbf{B K} & \mathbf{B K} \\
0 & \mathbf{A}-\mathbf{L} \mathbf{C}
\end{array}\right]}_{\mathbf{A}_{c l}}\left[\begin{array}{l}
\mathbf{x}(t) \\
\mathbf{e}(t)
\end{array}\right] \Leftarrow \text { augmented equation }
$$

- This says that the dynamics of both the state vector of the closed loop systems after introducing feedback as well as the estimation error are determined by the eigenvalues of $\mathbf{A}_{c /}$
- Since this is a block upper diagonal matrix, its eigenvalues are given by

$$
\left|s \mathbf{I}-\mathbf{A}_{C l}\right|=|s \mathbf{I}-(\mathbf{A}-\mathbf{B} \mathbf{K})| \cdot|s \mathbf{I}-(\mathbf{A}-\mathbf{L} \mathbf{C})|
$$

- This means that the poles of the closed loop system are the union of the regulator and estimator poles


## Combining regulator with state estimator

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$$
\begin{gathered}
{\left[\begin{array}{l}
\dot{\mathbf{x}}(t) \\
\dot{\mathbf{e}}(t)
\end{array}\right]=\underbrace{\left[\begin{array}{cc}
\mathbf{A}-\mathbf{B K} & \mathbf{B K} \\
0 & \mathbf{A}-\mathbf{L C}
\end{array}\right]}_{\mathbf{A}_{c l}}\left[\begin{array}{l}
\mathbf{x}(t) \\
\mathbf{e}(t)
\end{array}\right] \Leftarrow \text { augmented equation }} \\
\left|\left|s \mathbf{I}-\mathbf{A}_{c l}\right|=|s \mathbf{I}-(\mathbf{A}-\mathbf{B} \mathbf{K})| \cdot\right| s \mathbf{I}-(\mathbf{A}-\mathbf{L} \mathbf{C}) \mid
\end{gathered}
$$

- This also means that you can design the compensator and estimator separately just as we did before and then combine them (separation principle)
- As a design rule, you should place the estimator poles at >2 the real part of the regulator poles (found from transient specs required) to ensure the estimator converges fast and hence the estimated values used for feedback are good


## Combining regulator with state estimator

- The whole system (regulator + estimator) looks like


The full compensator
(Regulator + Estimator)

- Compensator accepts the sensor outputs as its inputs and provides at its output the actuator input
- Sometimes called dynamic output feedback compensator (DOFC)


## Full compensator equations



The full compensator
(Regulator + Estimator)

$$
\begin{aligned}
\dot{\hat{\mathbf{x}}} & =(\mathbf{A}-\mathbf{L C}-\mathbf{B K}) \hat{\mathbf{x}}+\mathbf{L} y \\
u & =-\mathbf{K} \hat{\mathbf{x}}
\end{aligned}
$$

$$
\mathbf{A}_{C} \equiv \mathbf{A}-\mathbf{L C}-\mathbf{B K}, \quad \mathbf{B}_{C} \equiv \mathbf{L}, \quad \mathbf{C}_{C} \equiv-\mathbf{K}
$$

TF of compensator $=-\mathbf{K}(\mathbf{s I}-(\mathbf{A}-\mathbf{L C}-\mathbf{B K}))^{-1} \mathbf{L}$

## Guidelines of designing full compensator

- Applying the separation principle, we simply design the controller matrix $\mathbf{K}$ and the observer matrix $L$ separately each to satisfy a desired set of poles
- $\mathbf{K}$ is found from the desired poles of closes loop system which can be obtained from the required specifications of the closed loop system (Max overshoot, settling time,...) similar to Lecture 9
- $L$ is found from the desired poles estimator which are either given or chosen to be 2-10 times larger than the system poles on the real axis
- Do not expect the feedback of the estimated state vector $u=-\mathbf{K} \hat{\mathbf{x}}$ to operate just as good as $u=-\mathbf{K x}$ especially at the initial transient period where the estimation error hasn't yet decayed from its initial nonzero value (refer to the MATLAB code of Lec. 12 for more insight)


## Pre-Scaling the full compensator (reference input)

- See section 11.6 page 857 of Dorf's book for more details



## Pre-Scaling the full compensator (reference input)

$$
\left.\begin{array}{rl}
\dot{\mathbf{x}}(t) & =\mathbf{A x}(t)+\mathbf{B} u(t) \quad \Leftarrow \text { plant state equation } \\
& =\mathbf{A x}(t)-\mathbf{B K} \hat{\mathbf{x}}(t)+\mathbf{B} N r \quad \Leftarrow \\
& =\mathbf{A x}(t)-\mathbf{B K}(\mathbf{x}(t)-\mathbf{e}(t))+\mathbf{B} N r \\
& =(\mathbf{A}-\mathbf{B K}) \mathbf{x}(t)+\mathbf{B K e}(t)+\mathbf{B} N r \\
\text { reference i} / \mathrm{p} u=N r-\mathbf{K} \hat{\mathbf{x}}
\end{array}\right) .
$$

