

**MALVINO & BATES**

**Electronic  
PRINCIPLES**

**SEVENTH EDITION**



# AC Models



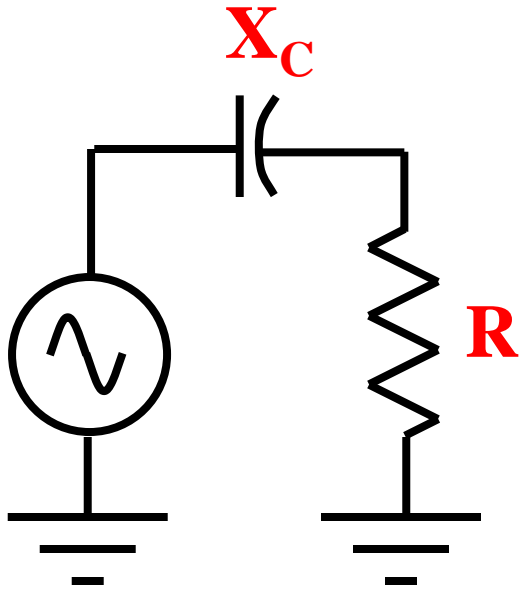
# Topics covered in Chapter 9

- **Base-biased amplifier**
- **Emitter-biased amplifier**
- **Small-signal operation**
- **AC beta**
- **AC resistance of the emitter diode**
- **Two transistor models**
- **Analyzing an amplifier**
- **AC quantities on the data sheet**

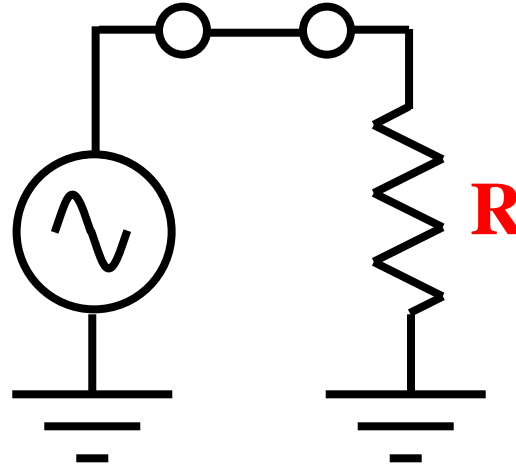
# Base-biased amplifier

- The **reactance** of a coupling capacitor is much smaller than the resistance
- **AC** input into base
- Amplified and inverted **output** at the collector
- **AC** output coupled to the load

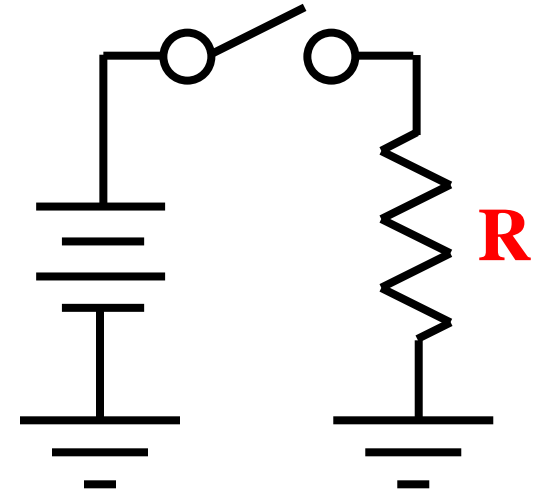
# The coupling capacitor



Good coupling:  
 $X_C < 0.1 R$

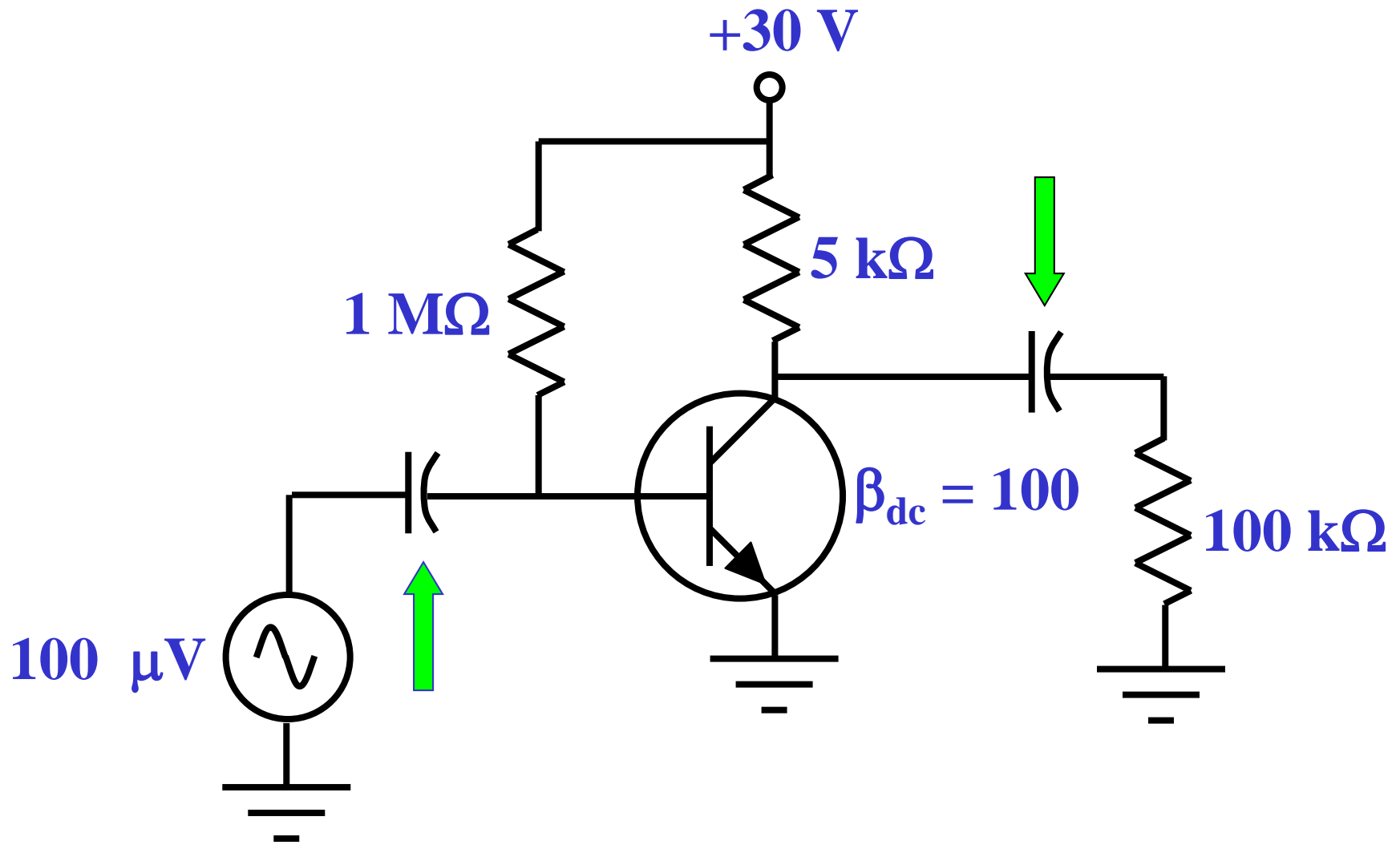


SHORT



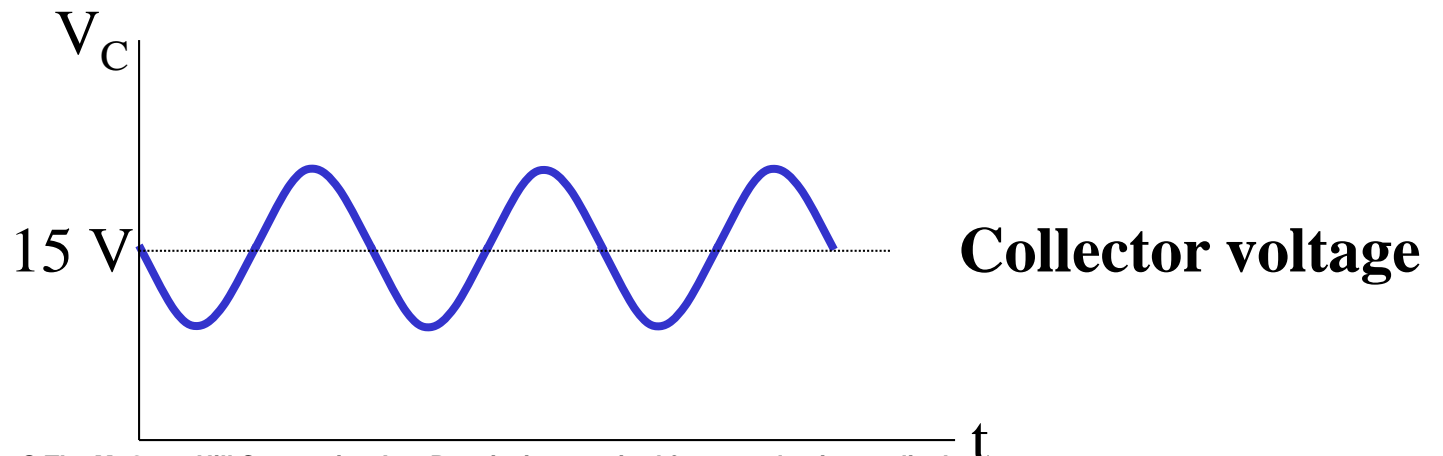
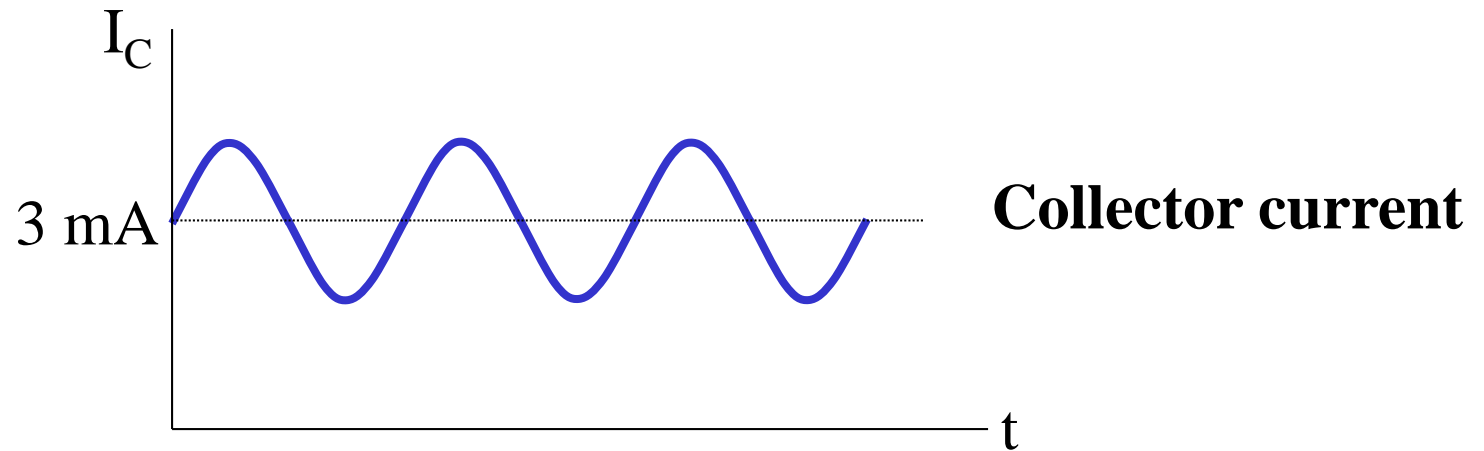
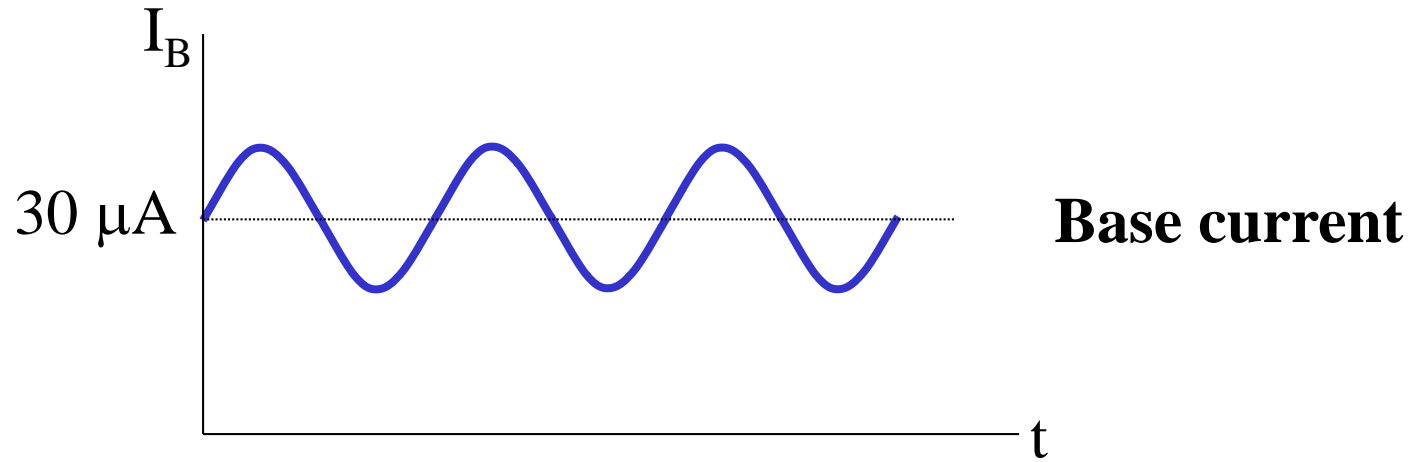
OPEN

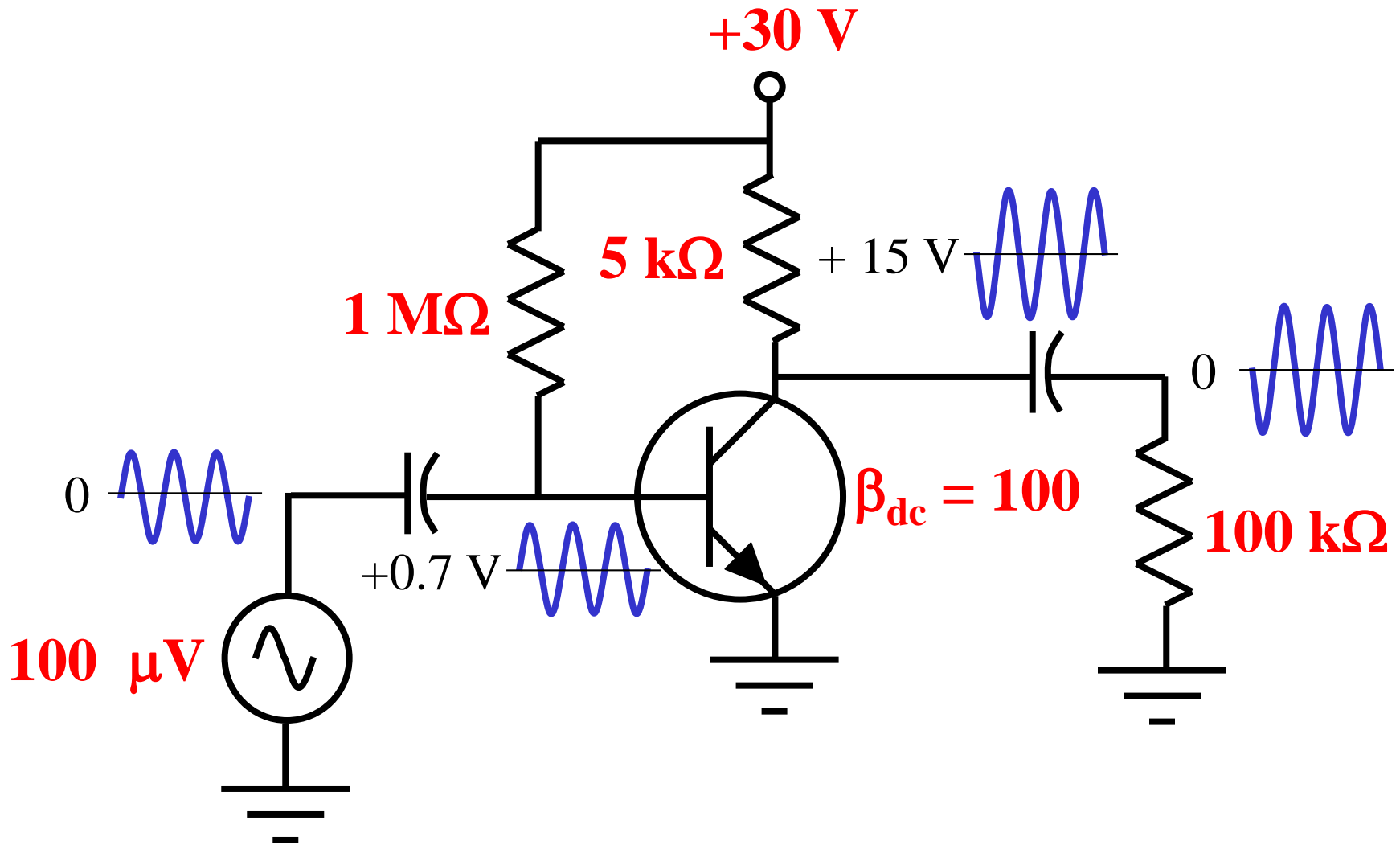
1. For **ac** analysis, the capacitor is a **short**.
2. For **dc** analysis, the capacitor is **open**.



A base-biased amplifier with **capacitive** coupling

A dc analysis reveals  $I_B = 30 \mu\text{A}$ ,  $I_C = 3 \text{ mA}$  and  $V_C = 15 \text{ V}$ .

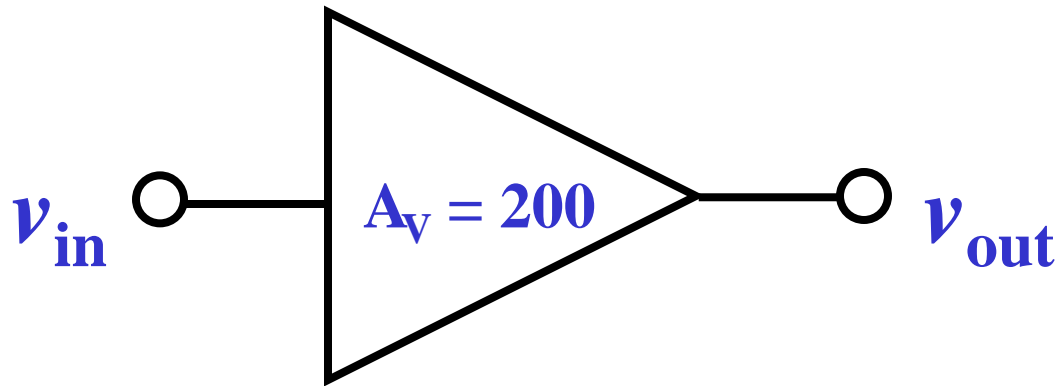




The base-biased amplifier with voltage waveforms

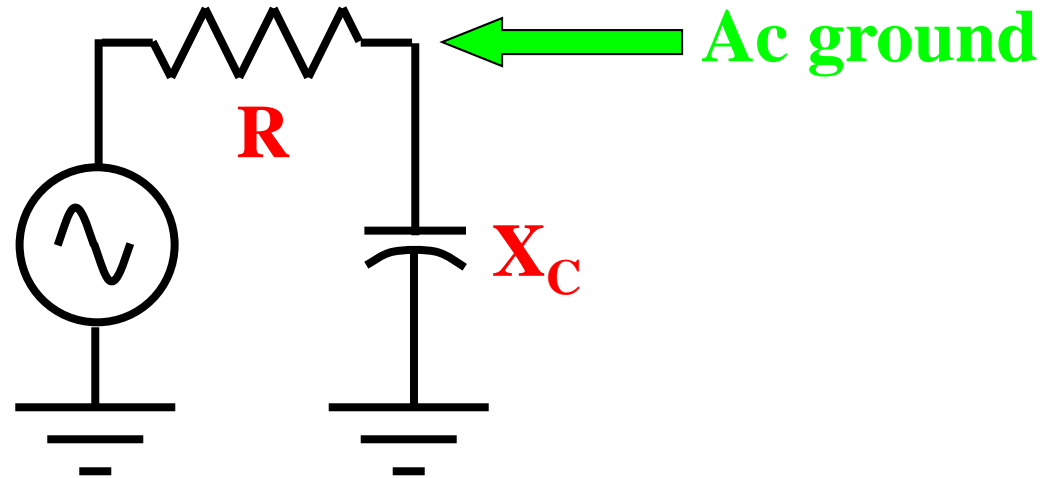


**The voltage gain of an amplifier is the ac output divided by the ac input.**



$$A_v = \frac{v_{out}}{v_{in}}$$

# The bypass capacitor



**Good bypassing:**

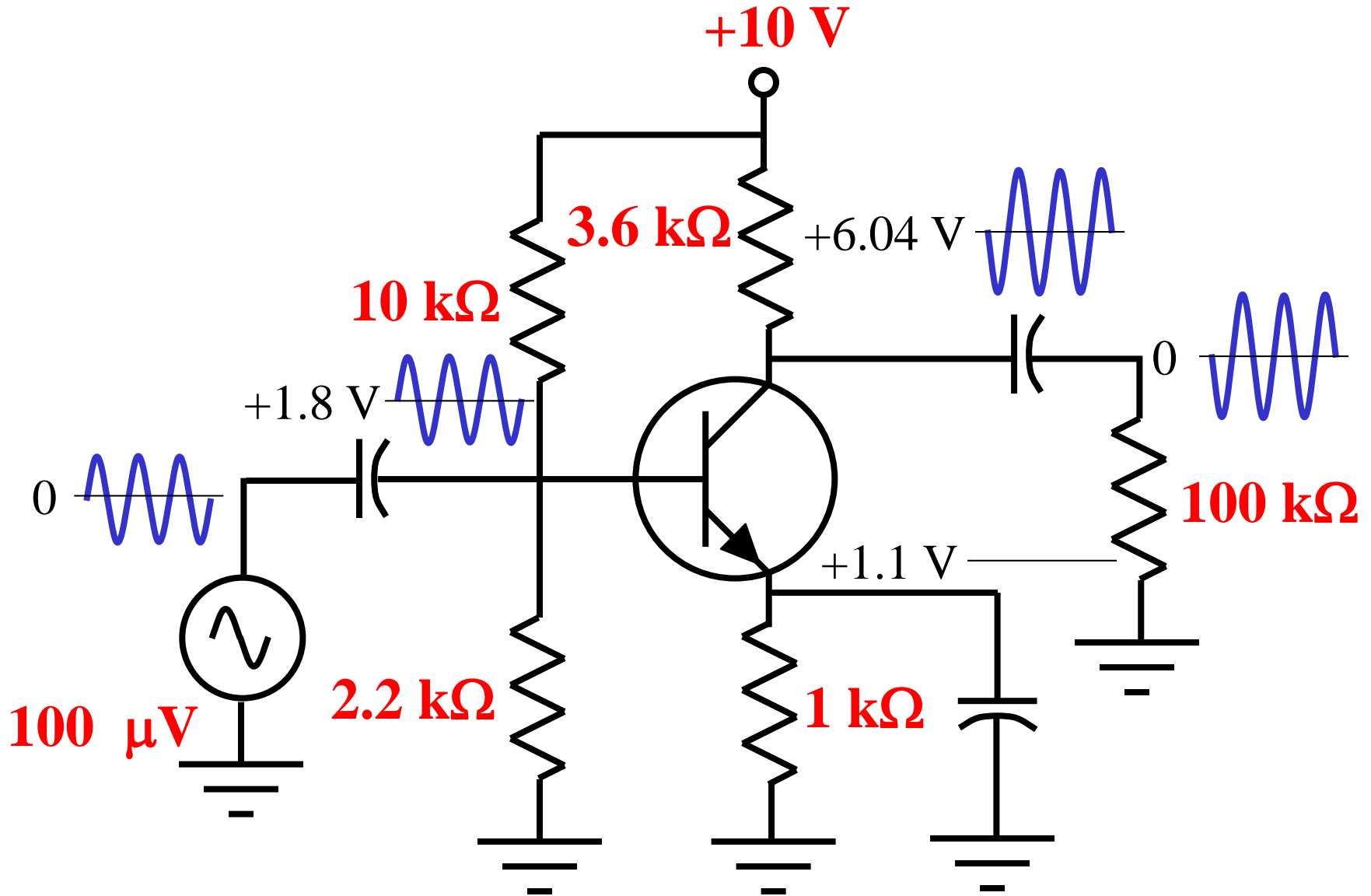
$$X_C < 0.1 R$$

**Note: The bypass capacitor appears open to dc and shorted to ac**

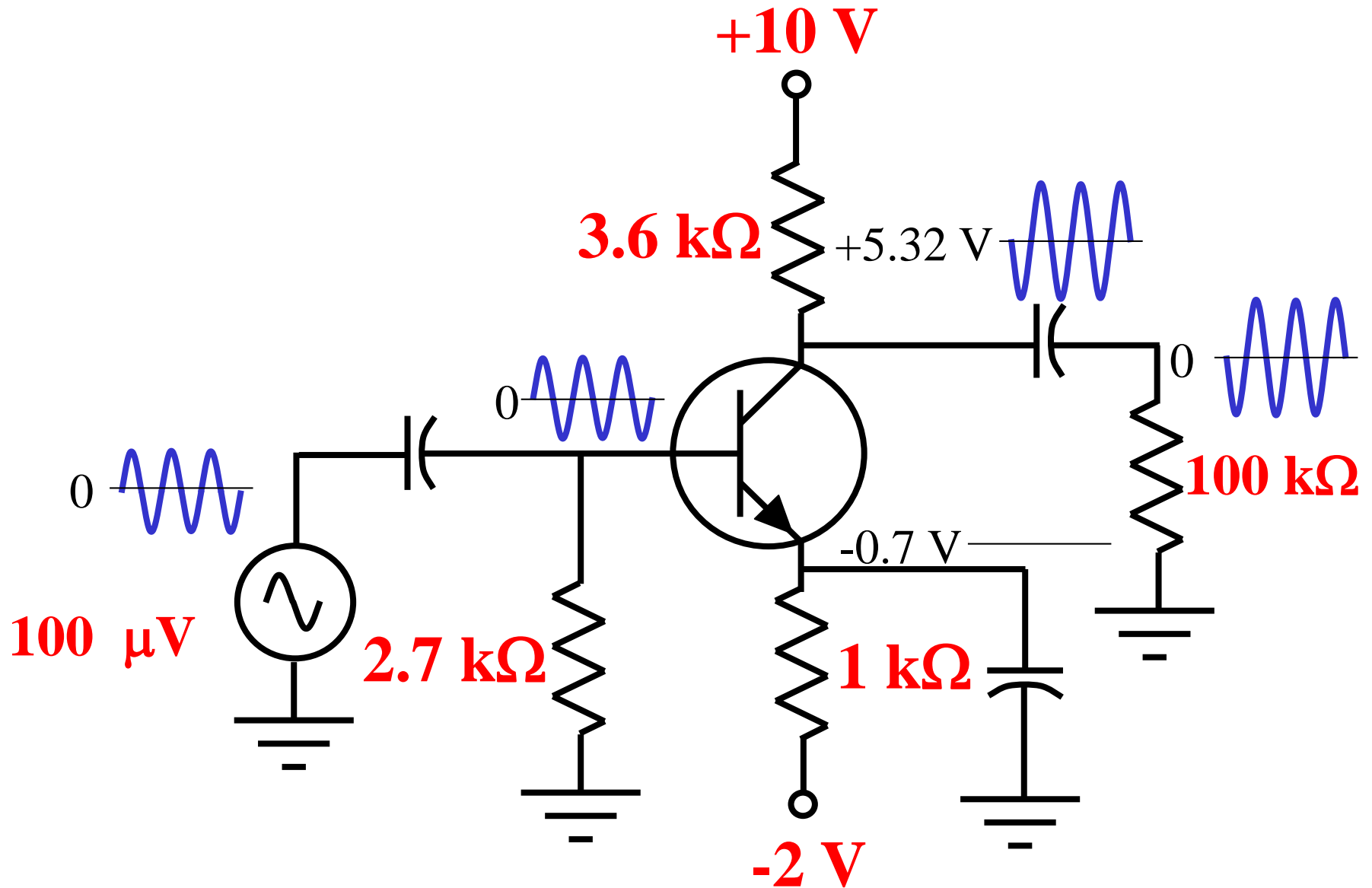
# VDB and TSEB amplifiers

- **Dc** voltages and currents are calculated mentally by opening capacitors
- The **ac** signal is coupled via a coupling capacitor
- The **bypass** capacitor causes an ac signal to appear across the base-emitter junction and provides higher gain

# A VDB amplifier with voltage waveforms

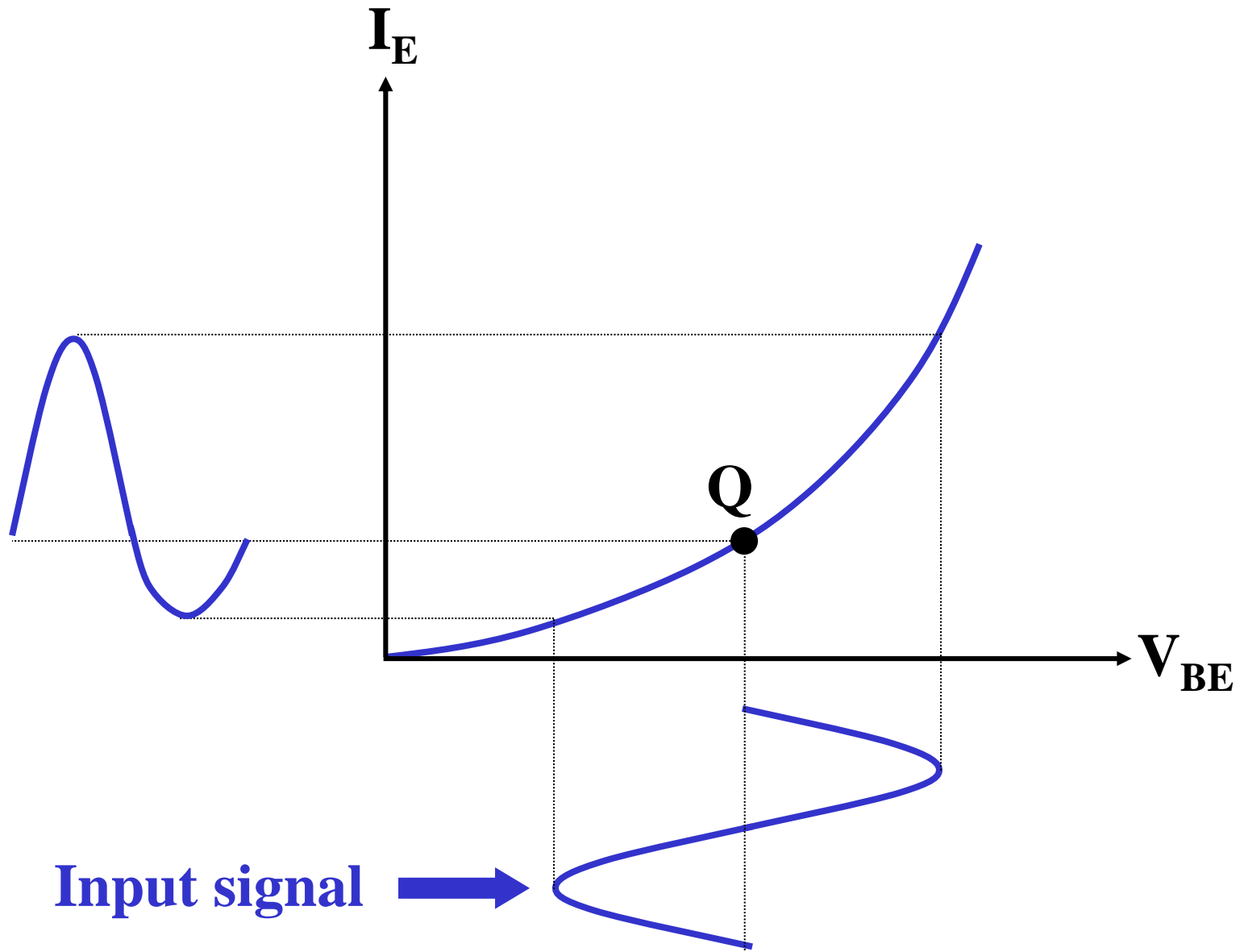


# A TSEB amplifier with voltage waveforms



# Distortion

- The **stretching and compressing** of alternate half cycles
- **Undesirable** in high-fidelity amplifiers
- Can be **minimized** by keeping the ac input small

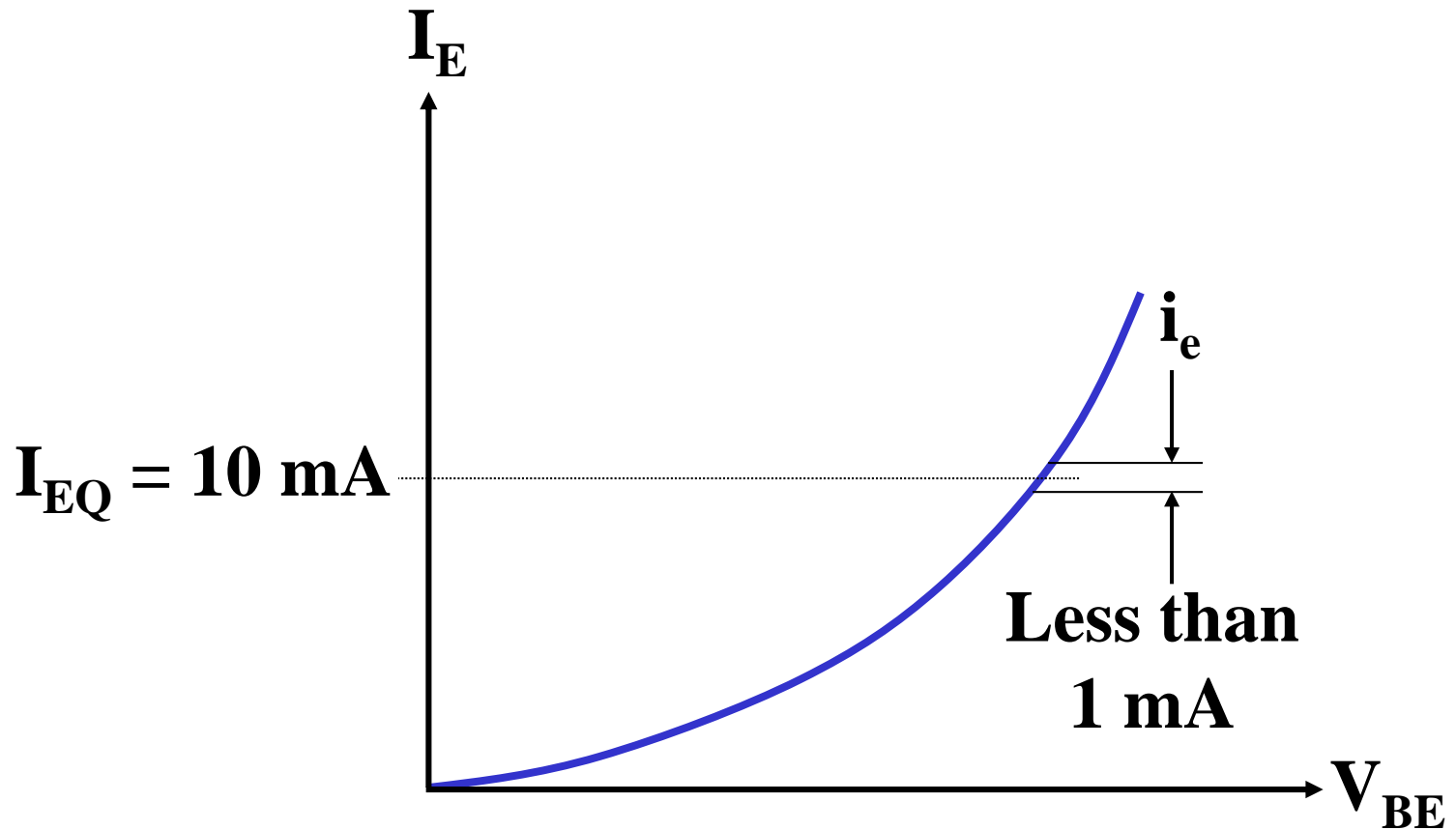


**Large-signal operation produces distortion**

# The 10 percent rule

- Total emitter current consists of dc and ac
- To **minimize** distortion,  $i_e$  must be small compared to  $I_{EQ}$
- The **ac** signal is small when the peak-to-peak ac emitter current is less than **10 percent** of the dc emitter current





**Total emitter current:  $I_E = I_{EQ} + i_e$**

**Small-signal operation:  $i_{e(\text{PP})} < 0.1I_{EQ}$**

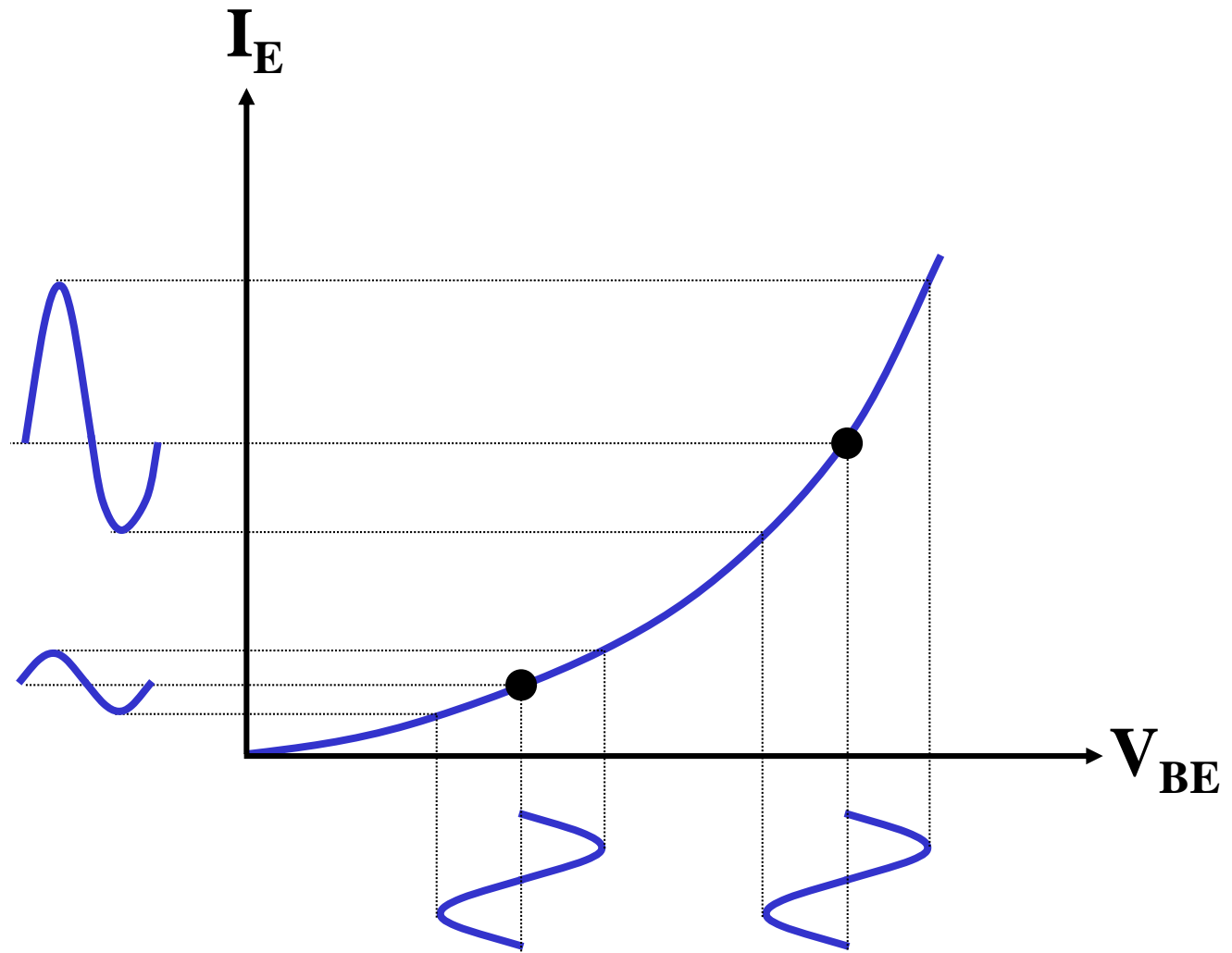
The dc current gain is given as:

$$\beta_{dc} = \frac{I_C}{I_B}$$

The ac current gain is given as:

$$\beta_{ac} = \frac{i_c}{i_b}$$

Use **CAPITAL** letters for dc quantities  
and **lowercase** letters for ac.



**The size of the ac emitter current depends on the Q point.**

**Total emitter current:**  $I_E = I_{EQ} + i_e$

**Total base-emitter voltage:**  $V_{BE} = V_{BEQ} + v_{be}$

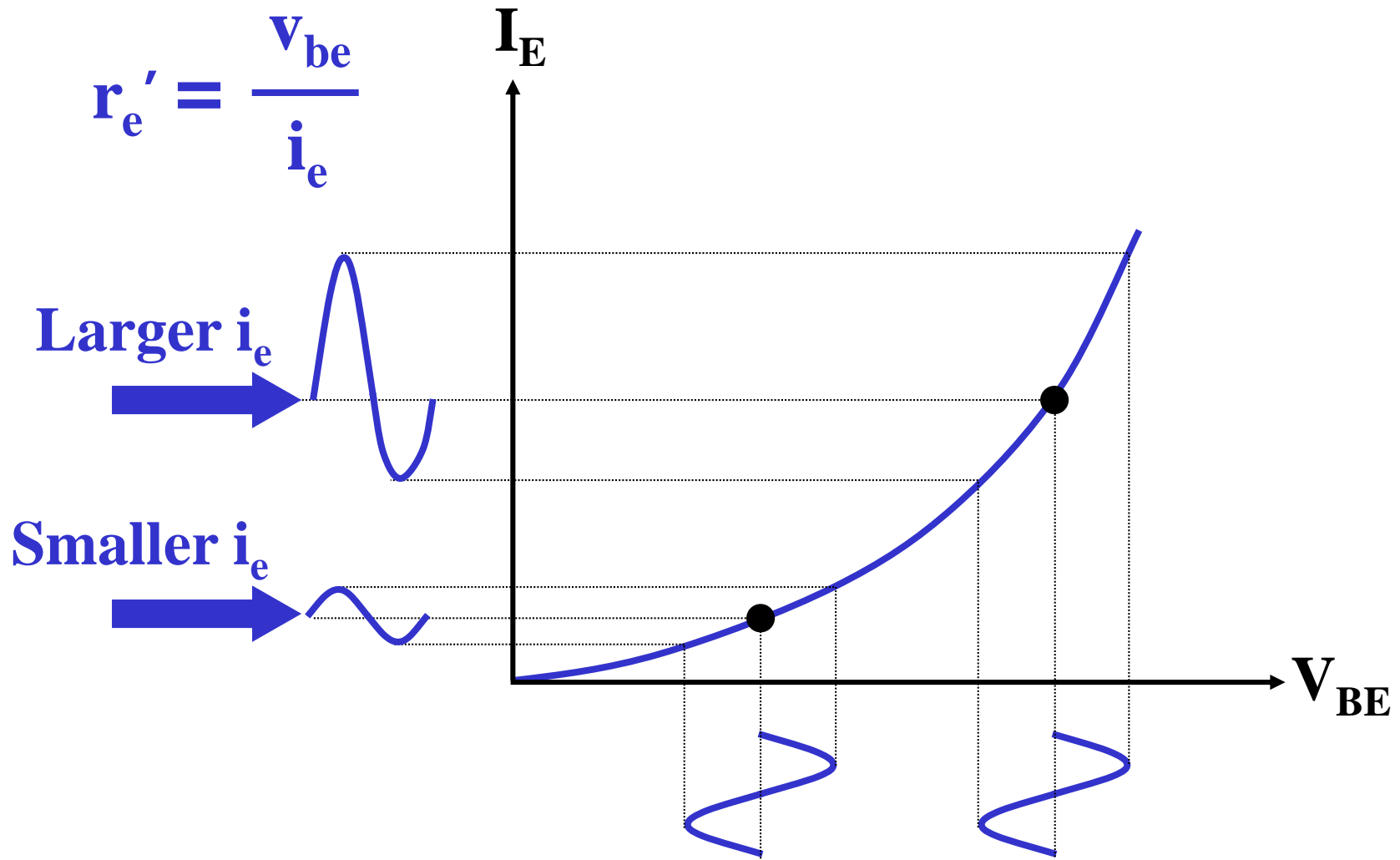
The **ac resistance** of the emitter diode is defined as:

$$r_e' = \frac{V_{be}}{i_e}$$

The **ac resistance** of the emitter diode **decreases** when the **dc** emitter current **increases**

# Ac resistance of the emitter diode

- Equals the **ac** base-emitter voltage divided by the **ac** emitter current
- The prime (') in  $r_e'$  indicates that the **resistance** is inside the transistor



**Note that  $r_e'$  varies with the operating point.**

**This implies that  $r_e'$  is a function of the dc emitter current.**

# Formula for ac emitter resistance

**Derived by using solid-state physics  
and calculus:**

$$r_e' = \frac{25 \text{ mV}}{I_E}$$

**Widely used in industry because of its  
simplicity and it applies to almost all  
commercial transistors**

# Transistor model

- **Ac equivalent circuit for a transistor**
- **Simulates** how a transistor behaves when an ac signal is present
- **Ebers-Moll (T model) and  $\pi$  type models are widely used**



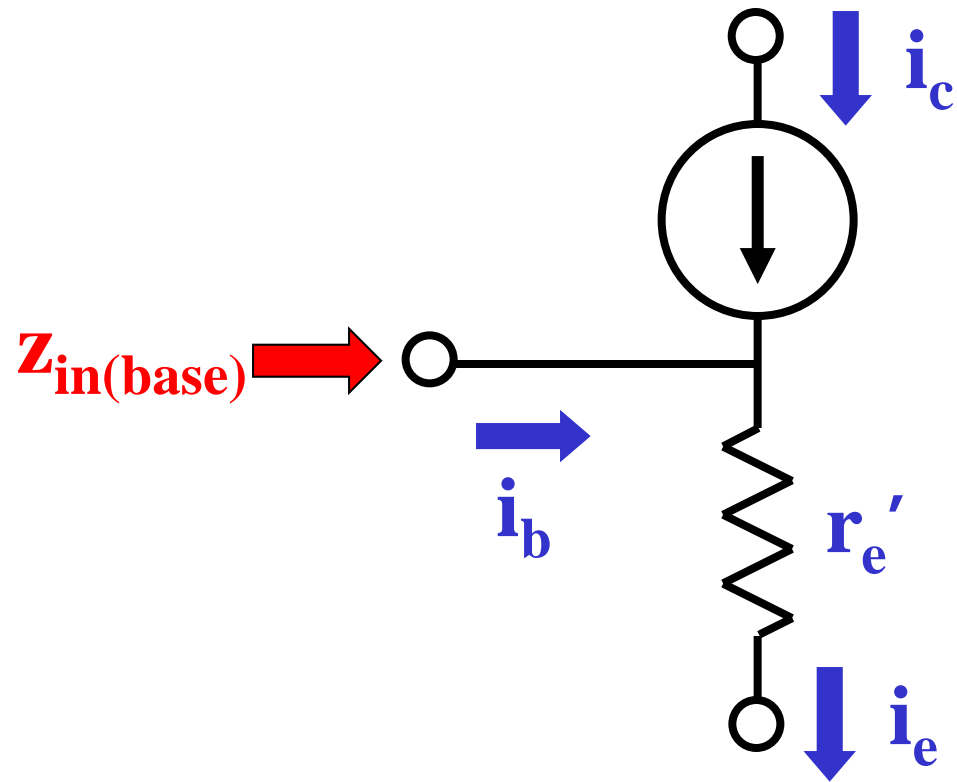
# The T model of a transistor:

$$Z_{\text{in(base)}} = \frac{V_{\text{be}}}{i_{\text{b}}}$$

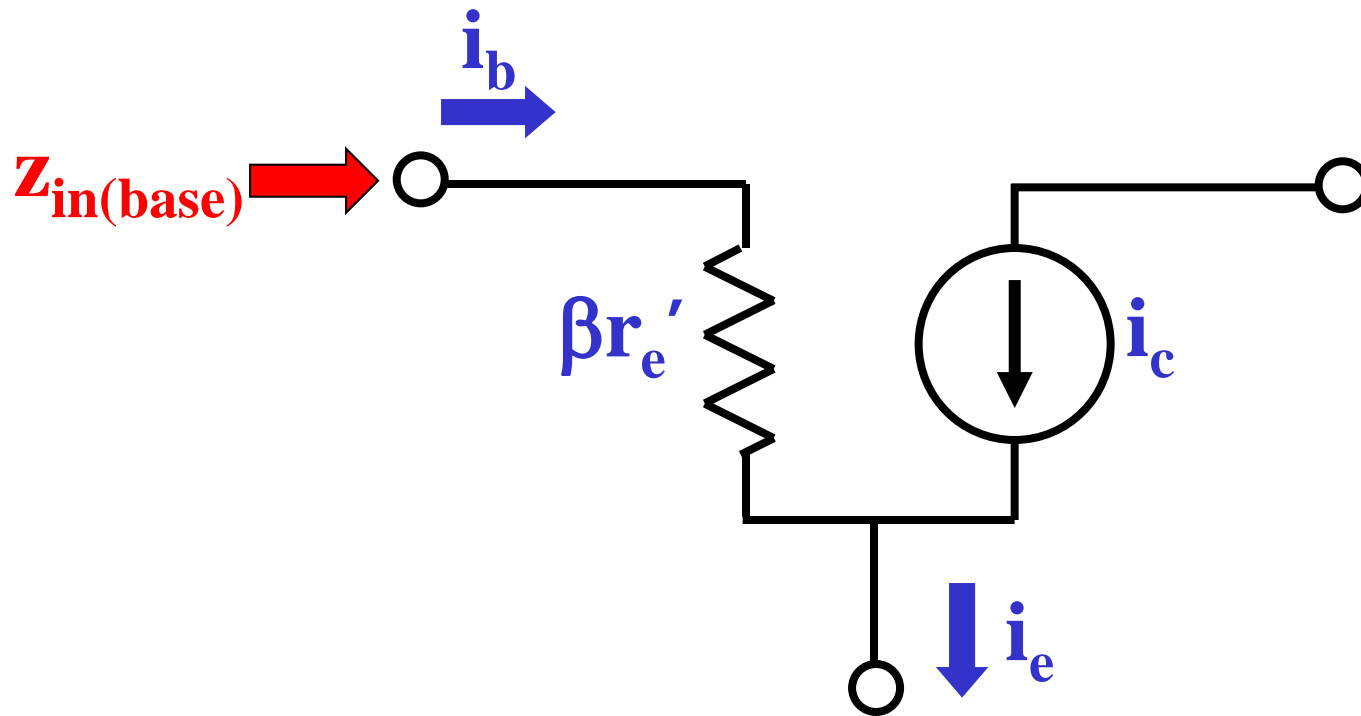
$$V_{\text{be}} = i_{\text{e}} r_{\text{e}}'$$

$$Z_{\text{in(base)}} = \frac{i_{\text{e}} r_{\text{e}}'}{i_{\text{b}}}$$

$$Z_{\text{in(base)}} = \beta r_{\text{e}}'$$



The  $\pi$  model of a transistor  
is based on  $Z_{in(base)} = \beta r_e'$  :

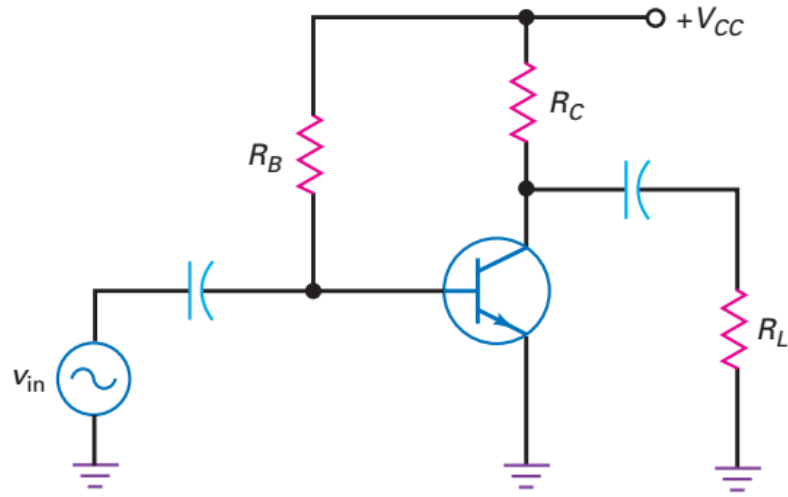


Clearly shows the **input impedance of  $\beta r_e'$**  will load the ac voltage source driving the base

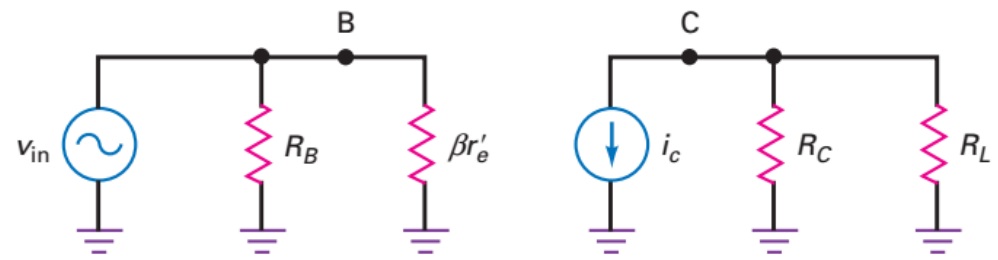
# Amplifier analysis

- **Perform** a complete dc analysis
- **Mentally** short all coupling and bypass capacitors for ac signals
- **Visualize** all dc supply voltages as ac grounds
- **Replace** the transistor by its  $\pi$  or T model
- **Draw** the ac equivalent circuit

# Base-Biased Amplifier and Its ac-Equivalent Circuit

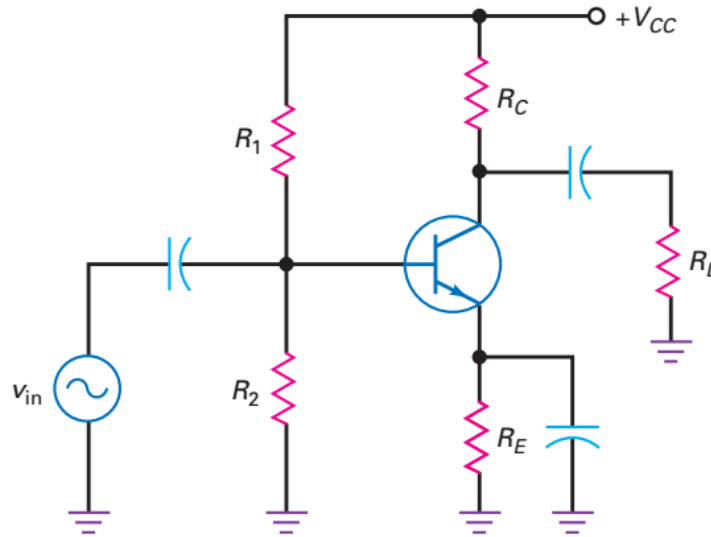


(a)

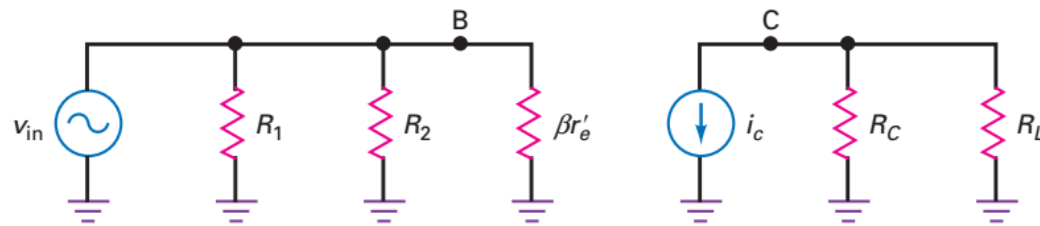


(b)

# VDB Amplifier and Its ac-Equivalent Circuit

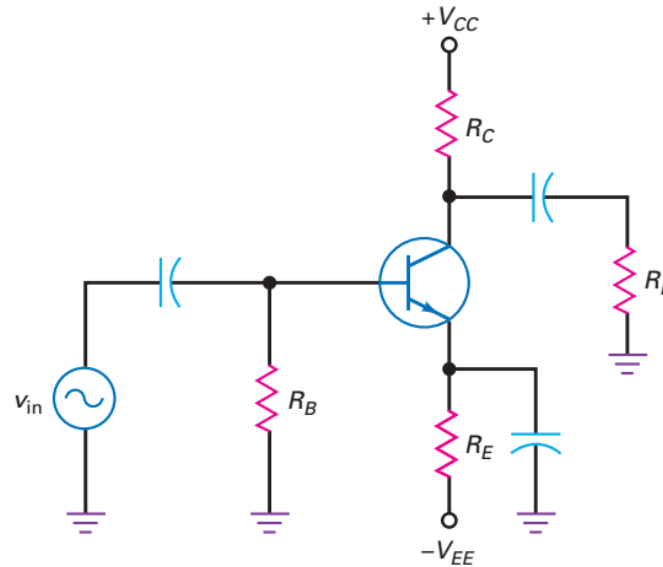


(a)

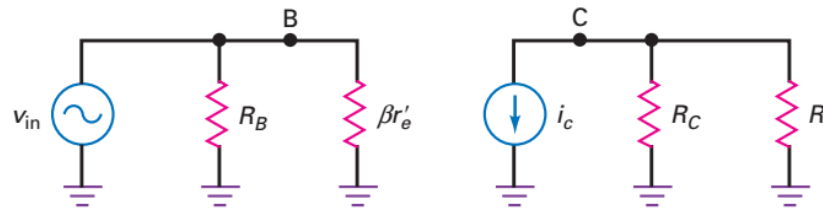


(b)

# TSEB Amplifier and Its ac-Equivalent Circuit

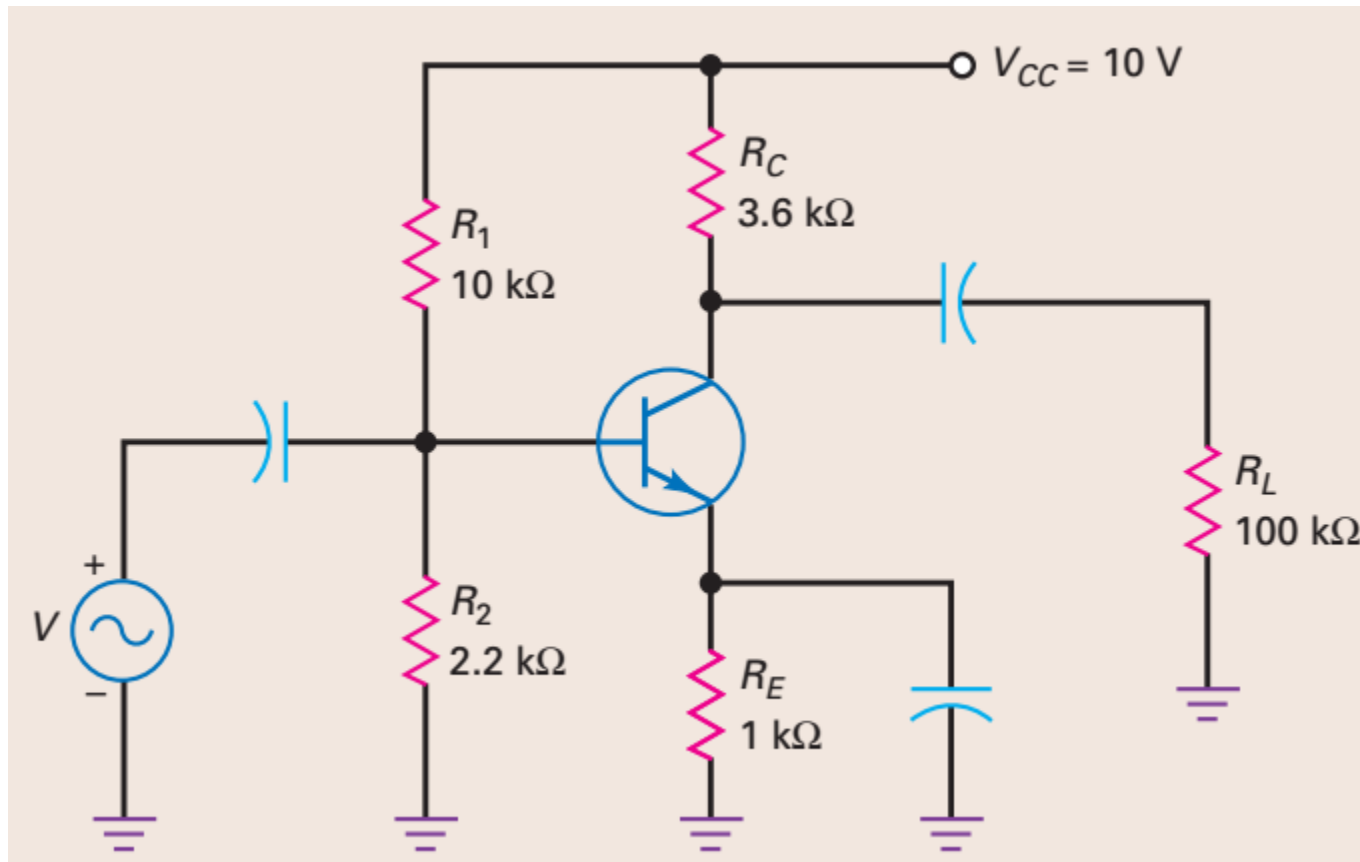


(a)

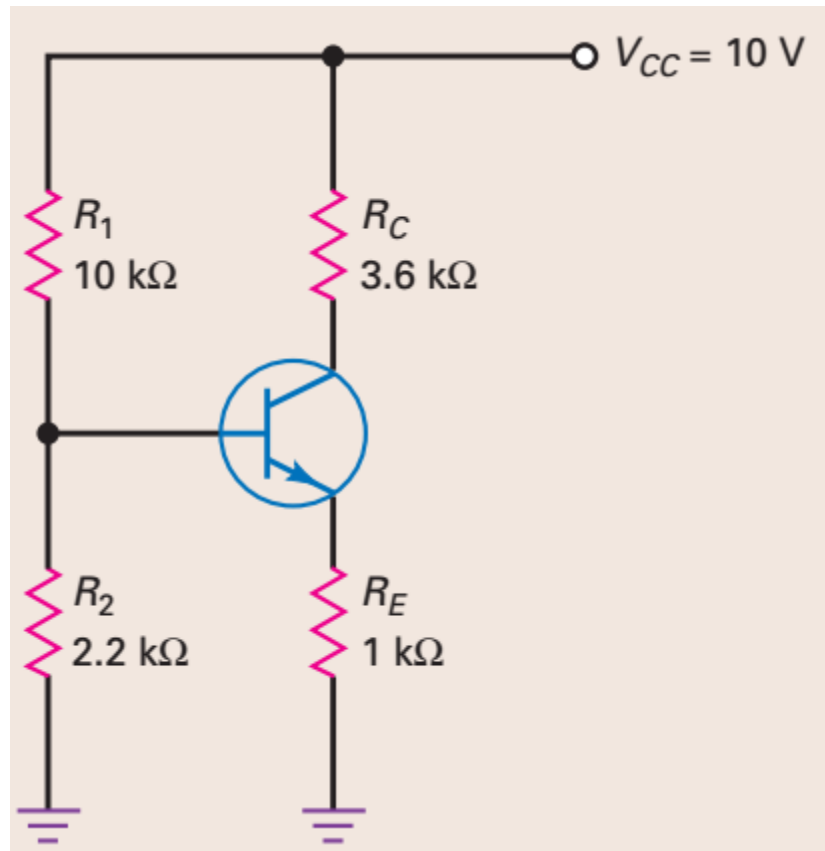


(b)

# Example: VDB DC and AC Equivalents



# Example: DC Equivalent



- Open all coupling and bypass capacitors.
- Redraw the circuit.
- Solve the dc circuit's Q point:

$$V_B = 1.8 \text{ V}$$

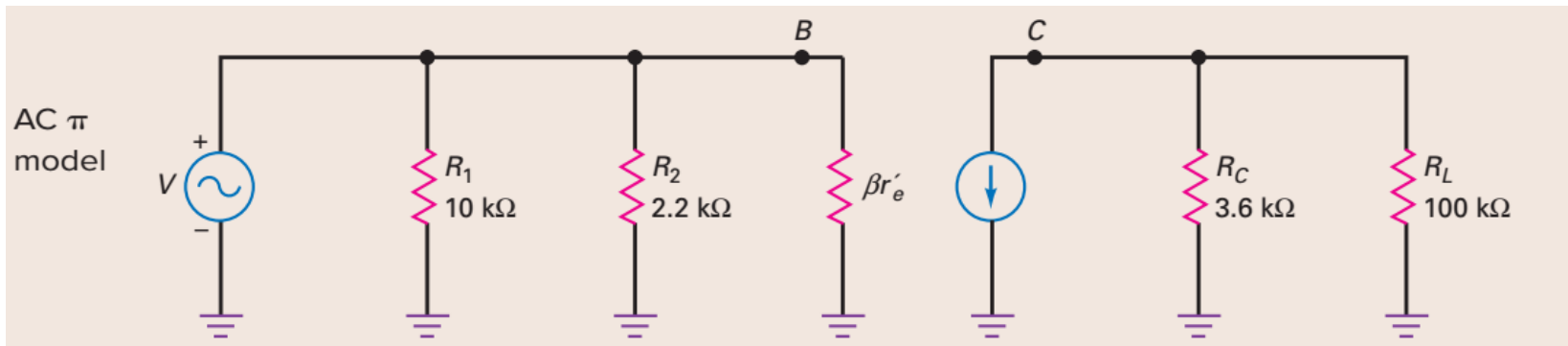
$$V_E = 1.1 \text{ V}$$

$$I_E = 1.1 \text{ mA}$$

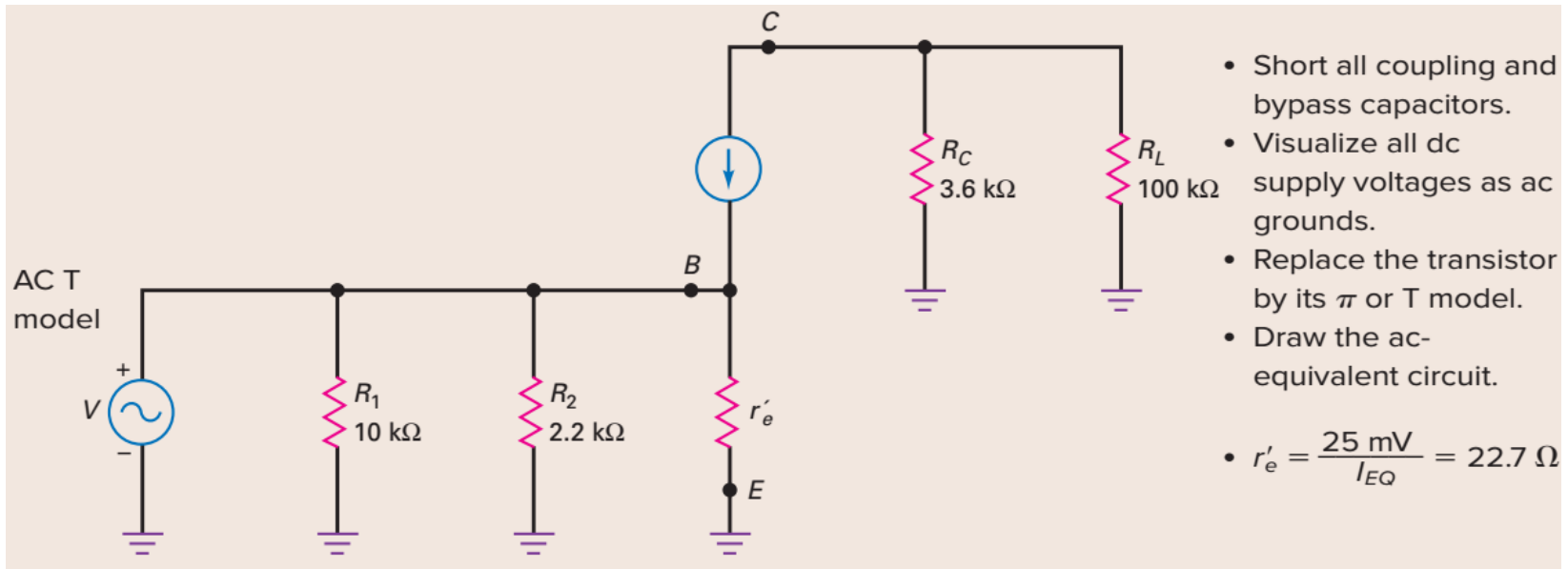
$$V_{CE} = 4.94 \text{ V}$$



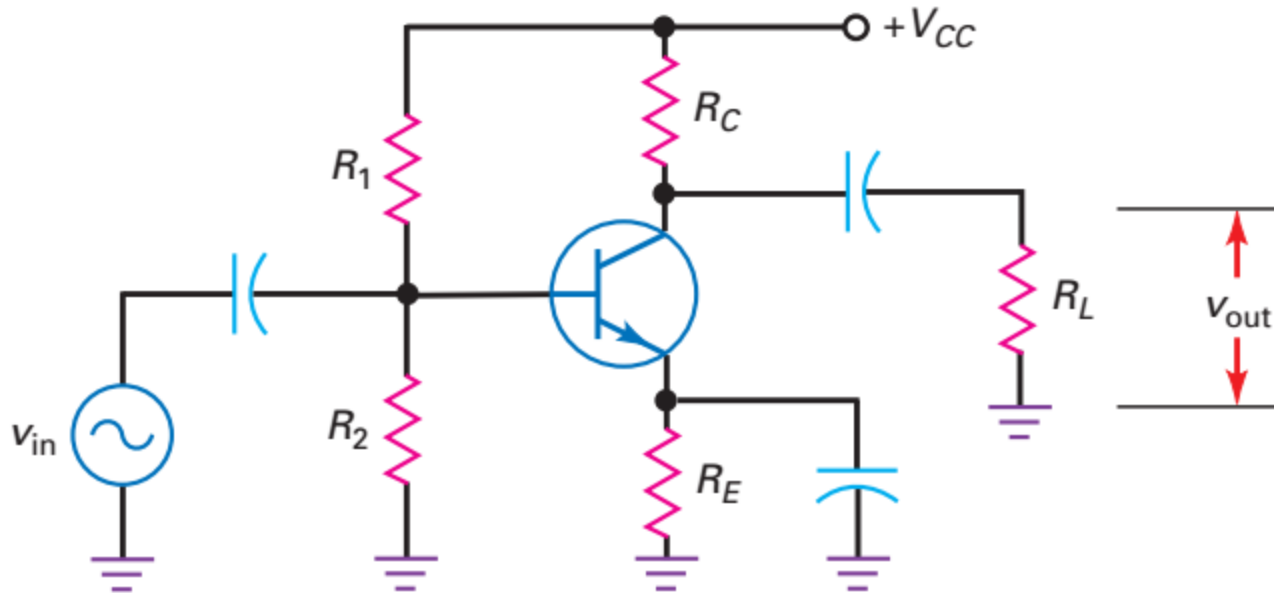
# Example: AC $\pi$ Model



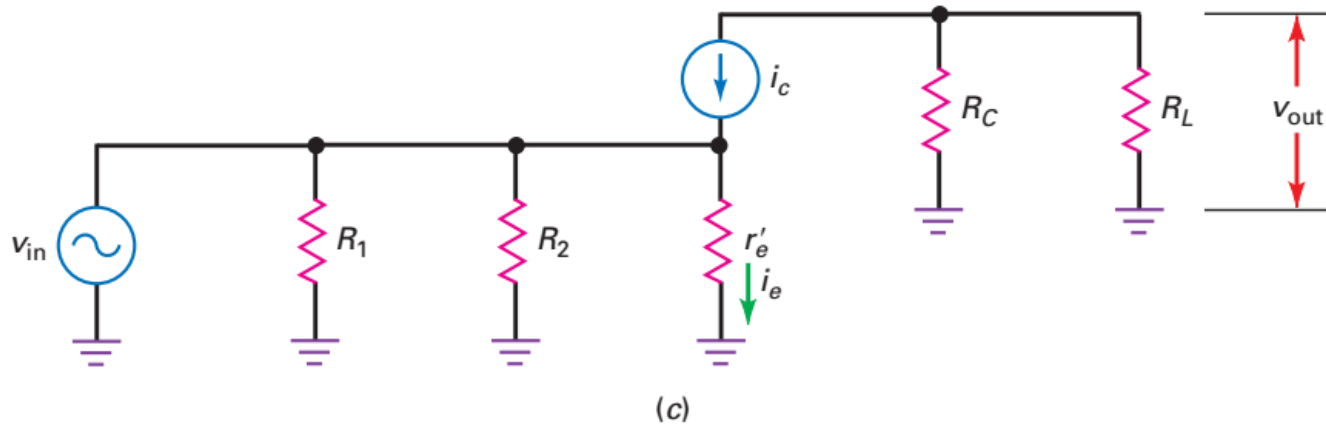
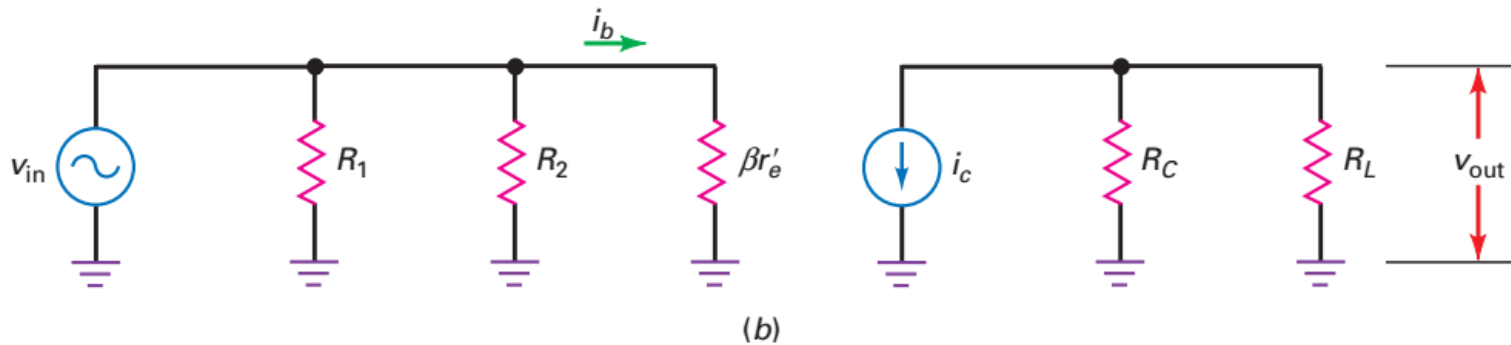
# Example: AC T Model



# Voltage Gain: VDB Amplifier



# AC Equivalent: $\pi$ and T Models



# Voltage Gain Calculation

## Derived from the $\pi$ Model

$$v_{\text{in}} = i_b \beta r'_e$$

$$v_{\text{out}} = i_c (R_C \parallel R_L) = \beta i_b (R_C \parallel R_L)$$

$$A_V = \frac{v_{\text{out}}}{v_{\text{in}}} = \frac{\beta i_b (R_C \parallel R_L)}{i_b \beta r'_e}$$

$$A_V = \frac{(R_C \parallel R_L)}{r'_e}$$

## Derived from the T Model

$$v_{\text{in}} = i_e r'_e$$

$$v_{\text{out}} = i_c r_c$$

$$r_c = R_C \parallel R_L$$

$$A_V = \frac{v_{\text{out}}}{v_{\text{in}}} = \frac{i_c r_c}{i_e r'_e}$$

Since  $i_c \approx i_e$ ,

$$A_V = \frac{r_c}{r'_e}$$

# Data sheets

- The four  $h$  parameters are a mathematical approach
- $h_{fe}$  is the ac current gain
- $h_{ie}$  is equivalent to input impedance
- $\beta_{ac} = h_{fe}$
- $r_e' = h_{ie}/h_{fe}$
- $h_{re}$  and  $h_{oe}$  are not needed for basic design and troubleshooting
- The  $h$  parameters give useful information when translated into  $r'$  parameters